

Advanced Digital Signal Processing

高等數位訊號處理

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課程網頁： <http://djj.ee.ntu.edu.tw/ADSP.htm>

歡迎大家來修課，也歡迎有問題時隨時聯絡！

- 評分方式：

Basic: 15 scores

原則上每位同學都可以拿到 12 分以上，
另外，上課回答問題，每回答一次加1分

Homework: 60 scores (5 times, 每 3 週一次)

請自己寫，和同學內容極高度相同，將扣 70% 的分數
就算寫錯但好好寫也會給 40~95% 的分數，
遲交分數打 8 折，**不交不給分**。不知道如何寫，可用 E-mail 和我
聯絡，或於上課時發問
禁止 Ctrl-C Ctrl-V 的情形。

Term paper 25 scores

Term paper 25 scores

方式有四種

(1) 書面報告

10頁以上(不含封面)，中英文皆可，11或12的字體，題目可選擇和課程有關的**任何一個**主題。

格式和一般寫期刊論文或碩博士論文相同，包括 abstract, conclusion, 及 references，並且要分 sections，必要時有 subsections。References 的寫法, 可參照一般 IEEE 的論文的寫法

鼓勵**多做實驗及模擬**，**有創新更好**。

嚴禁 Ctrl-C Ctrl-V 的情形，否則扣 70% 的分數

(2) Tutorial (對既有領域做淺顯易懂的整理)

限十七個名額，和書面報告格式相同，但頁數限制為18頁以上(若為加強前人的 tutorial，則頁數為 $(2/3)N + 13$ 以上， N 為前人 tutorial 之頁數)，題目由老師指定，以**清楚且有系統**的介紹一個主題的基本概念和應用為要求，為上課內容的進一步探討和補充，[交 Word 檔](#)。

選擇這個項目的同學，學期成績加 4分

(3) 口頭報告

限四個名額，每個人 40分鐘，題目可選擇和課程有關的任何一個主題。口頭報告將於 5月1日(第 8 週)進行。有意願的同學，請儘早告知，以先登記的同學為優先。

口頭報告時，鼓勵大家提問（包括口頭報告的同學，也可針對其他同學的報告內容提問）。曾經提問的同學，加分同上課回答問題。

選擇這個項目的同學，學期成績加 2分

(4) 編輯 Wikipedia

中文或英文網頁皆可，至少 2 個條目，但不可同一個條目翻成中文和英文。總計 80行以上。限和課程相關者，自由發揮，越有條理、有系統的越好

選擇編輯 Wikipedia 的同學，請於 6月26日前，向我登記並告知我要編輯的條目(2 個以上)，若有和其他同學選擇相同條目的情形，則較晚向我登記的同學將更換要編輯的條目

書面報告和編輯 Wikipedia，期限是 7月2日

Tutorial 可供選擇的題目(共 17 個，可以略做修改)

- (1) Guided Filter
- (2) Image Registration
- (3) Recent Development of Equalizer
- (4) Echo Cancellation
- (5) Rain Removal for Images
- (6) Distance Estimation for Sound Source
- (7) Signal Processing for Big Data
- (8) Fuzzy Logic in Signal Processing
- (9) Multiple Signal Classification (MUSIC) Algorithm
- (10) Image Stitching

Tutorial 可供選擇的題目(可以略做修改)

- (11) Photoacoustic Image Processing
- (12) Bioacoustics
- (13) Learning Based Image Compression Techniques
- (14) Shape Adaptive Image Compression
- (15) H.265 Video Compression Architecture
- (16) Anomaly Detection Using Neural Networks
- (17) Learning Based Prediction Techniques

上課時間：15 週

3/6,

3/13,

3/20, 出 HW1

3/27,

4/10, 交 HW1

4/17, 出 HW2

4/24,

5/1, Oral Presentation, 交 HW2

4/3 放假

5/8, 出 HW3

5/15,

5/22, 交 HW3

5/29, 出 HW4

6/5,

6/12, 交 HW4,

6/19, 出 HW5

7/2, 交 HW5 及 term paper

原則上: $3n$ 週出作業, $3n+2$ 週繳交

Matlab

Download: 請洽台大各系所

參考書目

洪維恩，Matlab 7 程式設計，旗標，台北市，2010.. (合適的入門書)

張智星，Matlab 程式設計入門篇，第三版，碁峰，2011.

蒙以正，數位信號處理：應用 Matlab，旗標，台北市，2007.

繆紹綱譯，數位影像處理：運用-Matlab，東華，2005.

預計看書學習所花時間：3~5 天

研究所和大學以前追求知識的方法有什麼不同？

研究所：觀念的學習

大學：

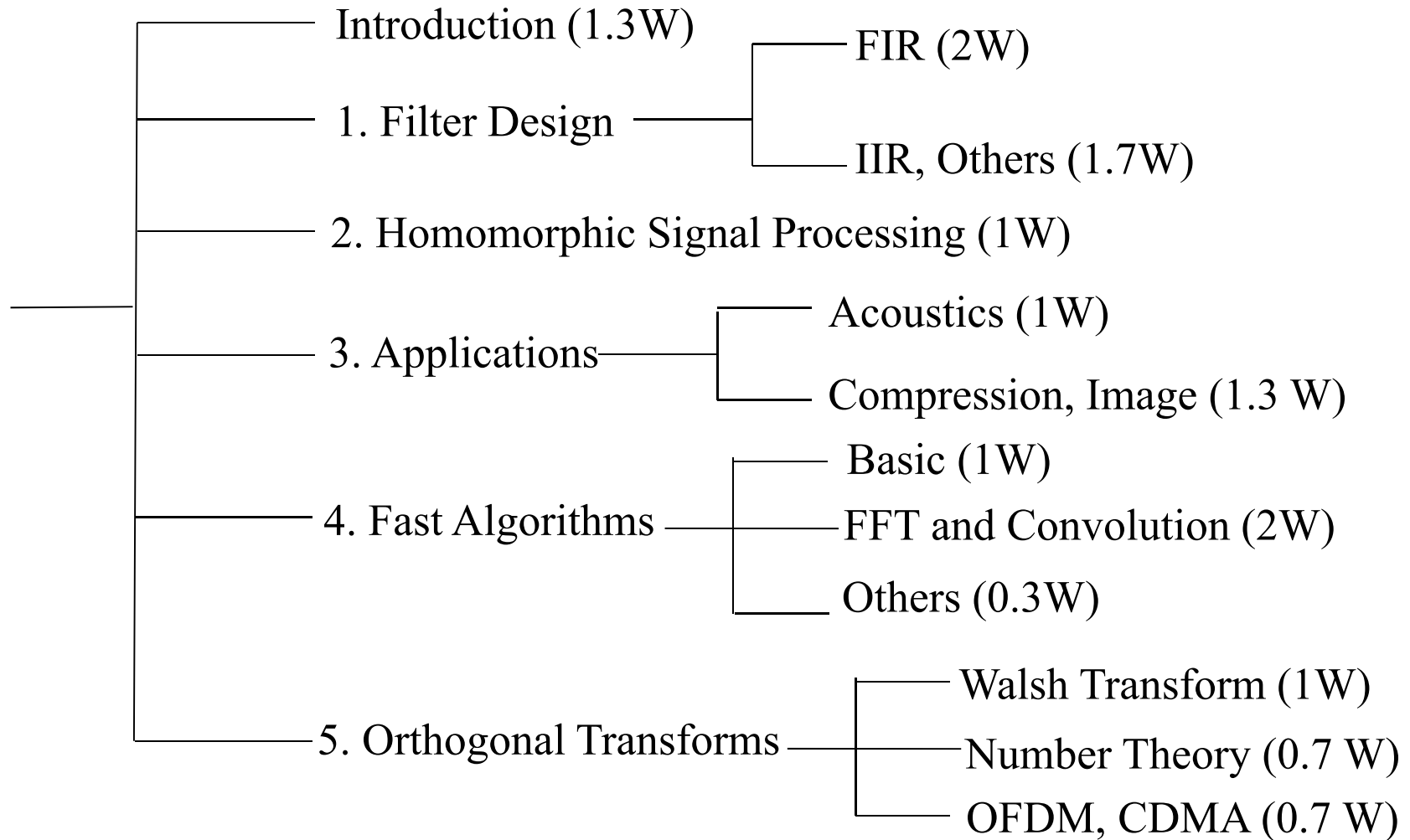
Question:

Why should we use the Fourier transform?

Is the Fourier transform the best choice in any condition?

I. Introduction

Outline

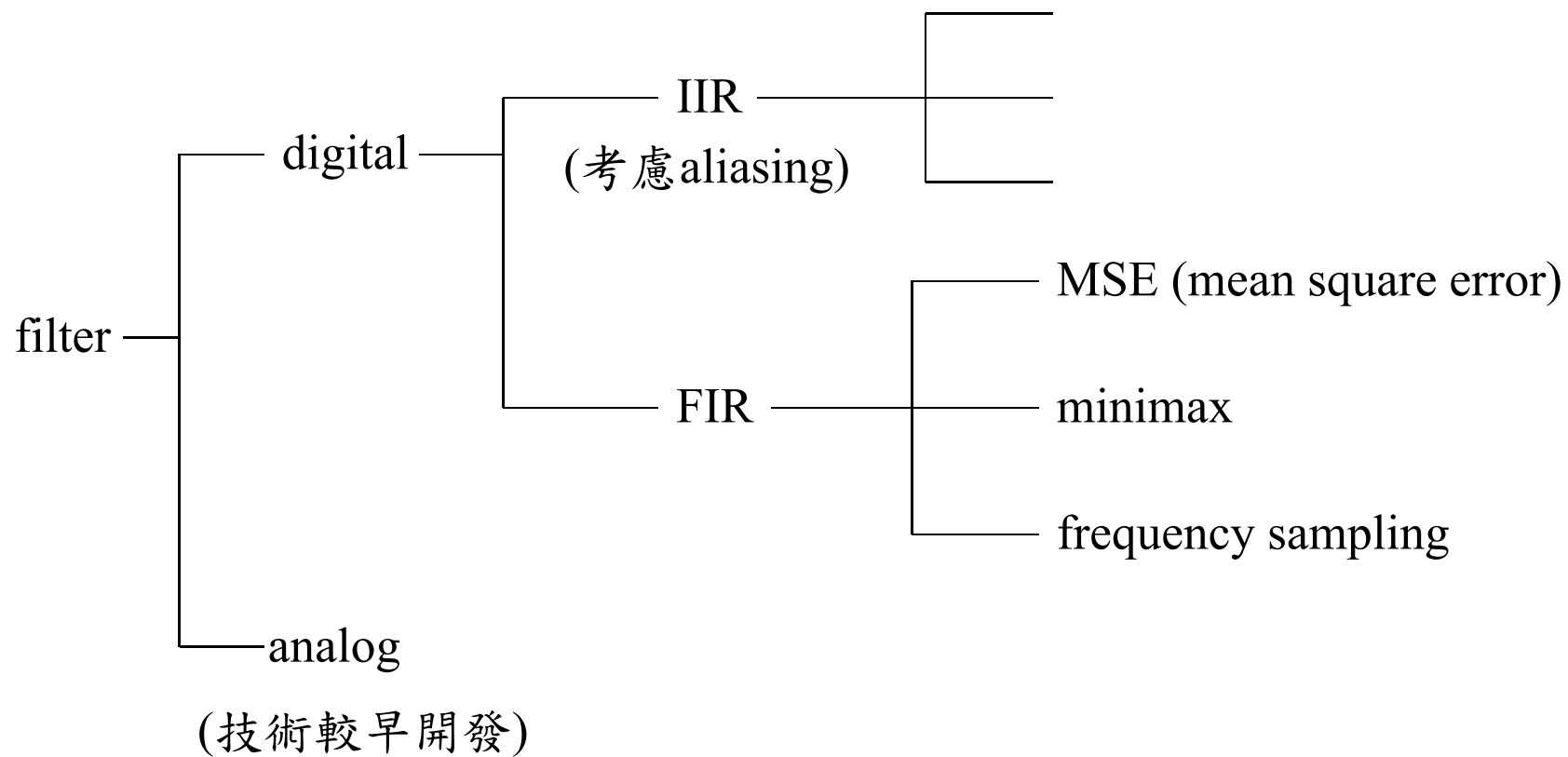


目標：

- (1) 對 Digital Signal Processing 作更有系統且深入的了解
- (2) 學習 Digital Signal Processing 幾個重要子領域的基礎知識

Part 1: Filter

- Filter 的分類



IIR filter 的優點：(1) easy to design

(2) (sometimes) easy to implement

缺點：

FIR filter 的優點：

缺點：An FIR filter is impossible to have the ideal frequency response of



Part 2: Homomorphic Signal Processing

- 概念：把 convolution 變成 addition

Part 3: Applications of DSP

filter design, data compression (image, video, text), acoustics (speech, music), image analysis (structural similarity, sharpness), 3D accelerometer

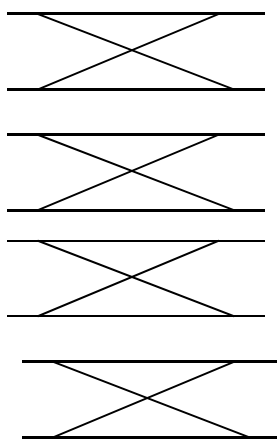
- **Part 4: Fast Algorithms**
- Basic Implementation Techniques

Example: one complex number multiplication
= ? Real number multiplication.

Trade-off: “Multiplication” takes longer than “addition”

- FFT and Convolution

Due to the Cooley-Tukey algorithm (butterflies),
the complexity of the FFT is:



The complexity of the convolution is: 3個 DFTs, $O(N \log_2 N)$

- **Part 5: Orthogonal Transforms**

DFT 的兩個主要用途:

Question: DFT 的缺點是什麼？ $DFT(x[n]) = \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi mn}{N}}$

- Walsh Transform
(CDMA)
- Number Theoretic Transform
- Orthogonal Frequency-Division Multiplexing (OFDM)
- Code Division Multiple Access (CDMA)

Review 1: Four Types of the Fourier Transform

(1) Fourier Transform

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt \quad , \quad x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi ft} df$$

Alternative definitions

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad , \quad x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

(2) Fourier series (suitable for period function)

$$X[m] = \int_0^T x(t) e^{-j\frac{2\pi m}{T}t} dt \quad x(t) = T^{-1} \sum_{m=-\infty}^{\infty} X[m] e^{j\frac{2\pi m}{T}t}$$

T : 週期 $x(t) = x(t+T)$

possible periods:

possible frequencies:

頻率和 m 之間的關係： $f = \frac{m}{T}$ $\frac{1}{T}$ 整數倍

(3) Discrete-time Fourier transform (DSP 常用)

$$X(f) = \sum_{n=-\infty}^{\infty} x[n] e^{-j2\pi f n \Delta_t}, \quad x[n] = \Delta_t \int_0^{1/\Delta_t} X(f) e^{j2\pi f n \Delta_t} df$$

$t = n\Delta_t$

Δ_t : sampling interval

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n \Delta_t}, \quad x[n] = \frac{\Delta_t}{2\pi} \int_0^{2\pi/\Delta_t} X(\omega) e^{j\omega n \Delta_t} d\omega$$

(4) Discrete Fourier transform (DFT) (DSP 常用)

$$X[m] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi mn}{N}}, \quad x[n] = \frac{1}{N} \sum_{m=0}^{N-1} X[m] e^{j\frac{2\pi mn}{N}}$$

頻率和 m 之間的關係： $f = \frac{m}{N\Delta_t} = \frac{m}{N} f_s$

where $f_s = 1/\Delta_t$ (sampling frequency)

- 四種 Fourier transforms 的比較

	time domain	frequency domain
(1) Fourier transform	continuous, aperiodic	continuous, aperiodic
(2) Fourier series	continuous, periodic (or continuous, only the value in a finite duration is known)	discrete, aperiodic
(3) discrete-time Fourier transform	discrete , aperiodic	continuous, periodic
(4) discrete Fourier transform	discrete, periodic (or discrete, only the value in a finite duration is known)	discrete, periodic

Review 2: Normalized Frequency

(1) Definition of **normalized frequency** F :

$$F = \frac{f}{f_s} = f \Delta_t = \frac{\omega \Delta_t}{2\pi} \quad \text{where } f_s = 1/\Delta_t \text{ (sampling frequency)}$$

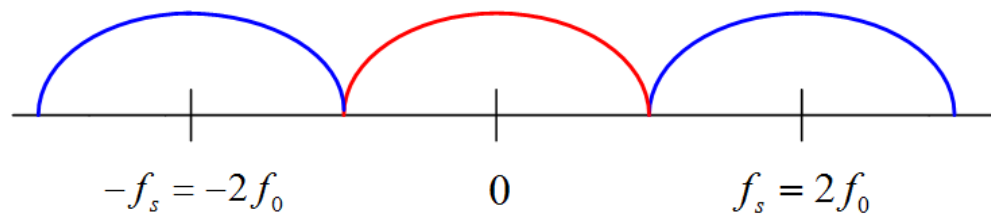
Δ_t : sampling interval

(2) folding frequency f_0

$$f_0 = \frac{f_s}{2} \quad \text{若以 normalized frequency 來表示,}$$

folding frequency = 1/2

$H(f)$:



For the discrete time Fourier transform

$$(1) G(f) = G(f + f_s) \longrightarrow \text{i.e., } G(F) = G(F + 1).$$

$$(2) \text{ If } g[n] \text{ is real } \longrightarrow G(F) = G^*(-F) \text{ (* means conjugation)}$$

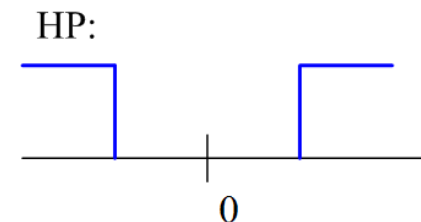
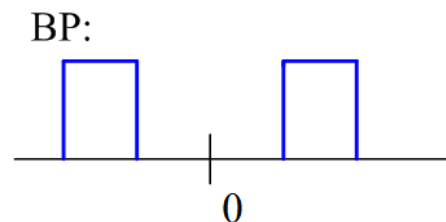
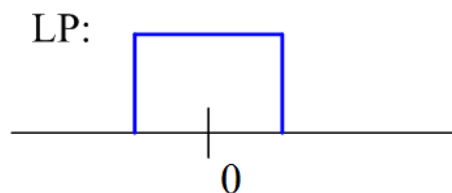
只需知道 $G(F)$ for $0 \leq F \leq \frac{1}{2}$ (即 $0 < f < f_0$)

就可以知道全部的 $G(F)$

$$(3) \text{ If } g[n] = g[-n] \text{ (even)} \longrightarrow G(F) = G(-F),$$

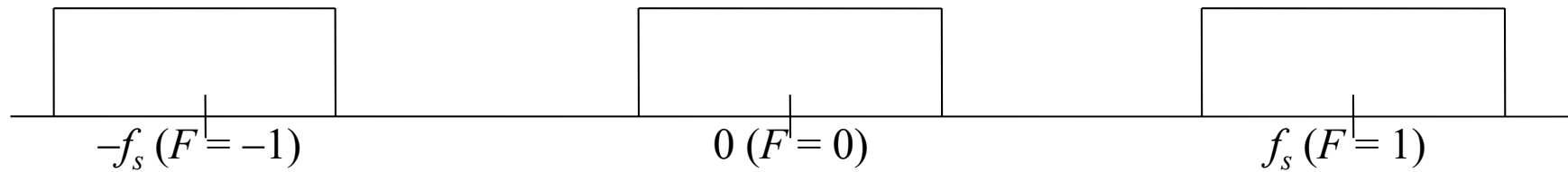
$$g[n] = -g[-n] \text{ (odd)} \longrightarrow G(F) = -G(-F)$$

Analog
filter: $H(f)$

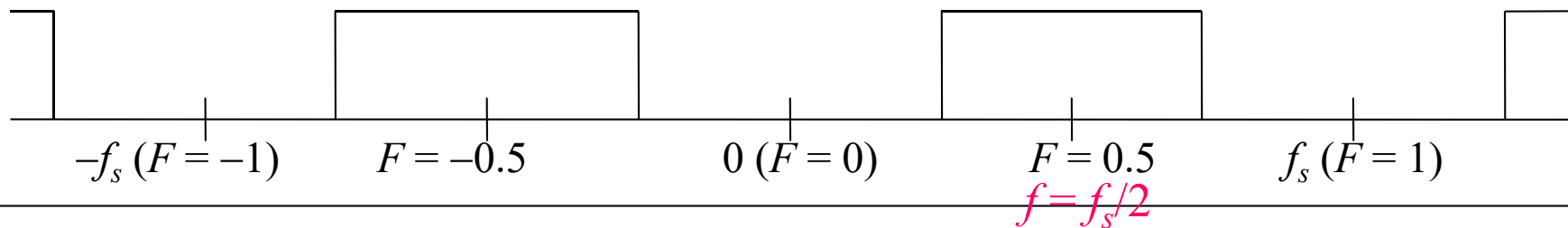


- Discrete time Fourier transform of the lowpass, highpass, and band pass filters

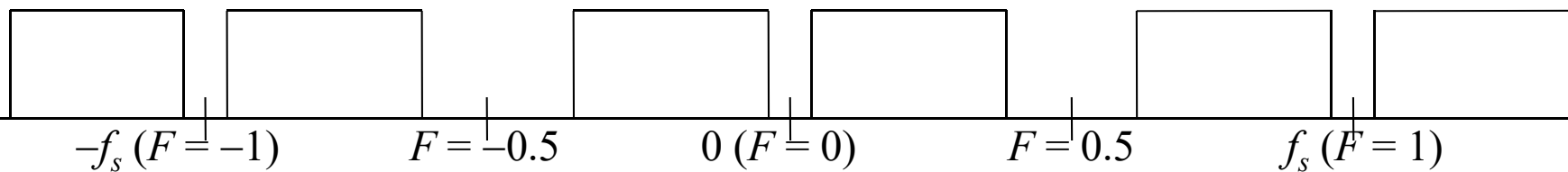
low pass filter (pass band 在 f_s 的整數倍附近)



high pass filter



band pass filter



Review 3: Z Transform and Laplace Transform

- **Z-Transform**

suitable for **discrete** signals

$$G(z) = \sum_{n=-\infty}^{\infty} g[n]z^{-n}$$

Compared with the discrete time Fourier transform:

$$G(f) = \sum_{n=-\infty}^{\infty} g[n]e^{-j2\pi f n\Delta_t} \quad z = e^{j2\pi f \Delta_t}$$

- **Laplace Transform**

suitable for **continuous** signals

One-sided form $G(s) = \int_0^{\infty} g(t)e^{-st} dt$

Two-sided form $G(s) = \int_{-\infty}^{\infty} g(t)e^{-st} dt$

Compared with the Fourier transform:

$$G(f) = \int_{-\infty}^{\infty} g(t)e^{-j2\pi ft} dt \quad s = j2\pi f$$

Review 4: IIR Filter Design

Two types of digital filter:

- (1) IIR filter (infinite impulse response filter)
- (2) FIR filter (finite impulse response filter)

There are 3 popular methods to design the IIR filter.

Method 1: Impulse Invariance

白話一點，就是直接做 sampling

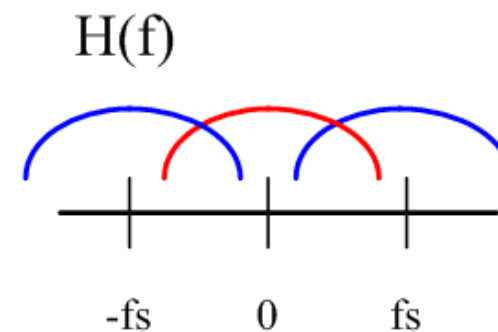
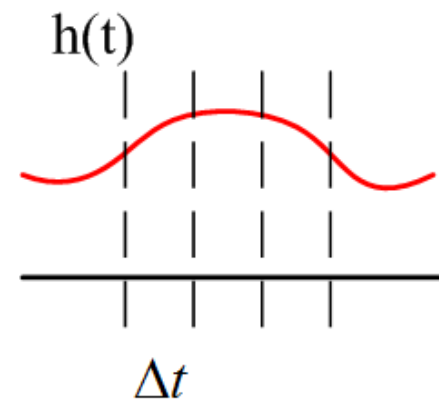
analog filter $h_a(t)$

digital filter $h[n]$

$$h[n] = h_a(n\Delta_t)$$

Advantage : Simple

Disadvantage : (1) infinite
(2)



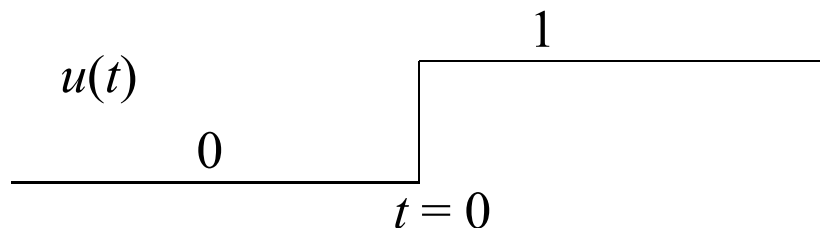
Method 2: Step Invariance

對 step function 的 response 作 sampling

analog filter $h_a(t)$

digital filter $h[n]$

step function (continuous form)



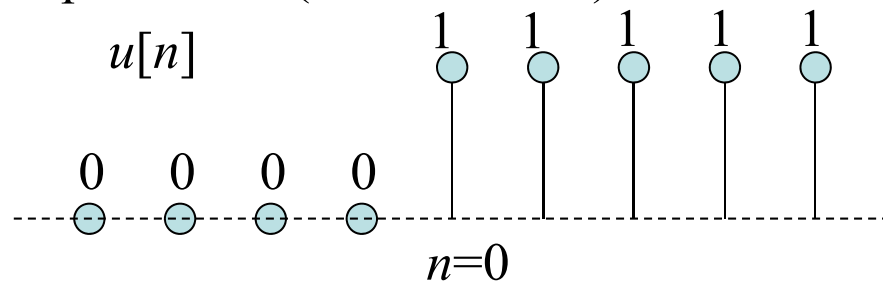
Laplace transform of $u(t)$:

$$\frac{1}{s}$$

Fourier transform of $u(t)$:

$$\frac{1}{j2\pi f}$$

step function (discrete form)



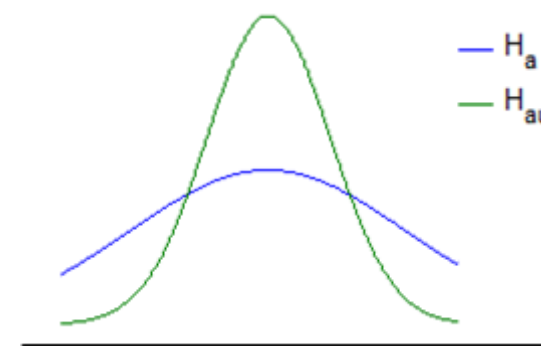
Z transform of $u[n]$:

$$\frac{1}{1 - z^{-1}}$$

Step 1 Calculate the convolution of $h_a(t)$ and $u(t)$

$$h_{a,u}(t) = h_a(t) * u(t) = \int_{-\infty}^{\infty} h_a(\tau)u(t-\tau)d\tau = \int_{-\infty}^t h_a(\tau)d\tau$$

$$H_{a,u}(f) = \frac{H_a(f)}{j2\pi f} \quad (\text{其實就是對 } h_a(t) \text{ 做積分})$$



Step 2 Perform sampling for $h_{a,u}(t)$

$$h_u[n] = h_{a,u}(n\Delta_t)$$

Step 3 Calculate $h[n]$ from $h[n] = h_u[n] - h_u[n-1]$

Note: Since $h_u[n] = h[n] * u[n]$ $H_u(z) = \frac{1}{1-z^{-1}} H(z)$

$$H(z) = (1-z^{-1})H_u(z)$$

so $h[n] = h_u[n] - h_u[n-1]$

Advantage of the step invariance method:

* 主要 Advantage:

Disadvantage of the step invariance method:

較為間接，設計上稍微複雜

Method 3: Bilinear Transform

Suppose that we have known an analog filter $h_a(t)$ whose frequency response is $H_a(f)$.

To design the digital filter $h[n]$ with the frequency response $H(f)$,

$$H(f_{new}) = H_a(f_{old})$$

$$f_{old} \in (-\infty, \infty)$$

$$f_{new} \in (-f_s/2, f_s/2)$$

$$f_s = 1/\Delta_t \text{ (sampling frequency)}$$

- The relation between f_{new} and f_{old} is determined by the mapping function

$$s = c \frac{1 - z^{-1}}{1 + z^{-1}}$$

s : index of the Laplace transform

z : index of the Z transform

c : some constant

$$s = c \frac{1 - z^{-1}}{1 + z^{-1}}$$

$$s = j2\pi f_{old}$$

$$z = e^{j2\pi f_{new} \Delta_t}$$

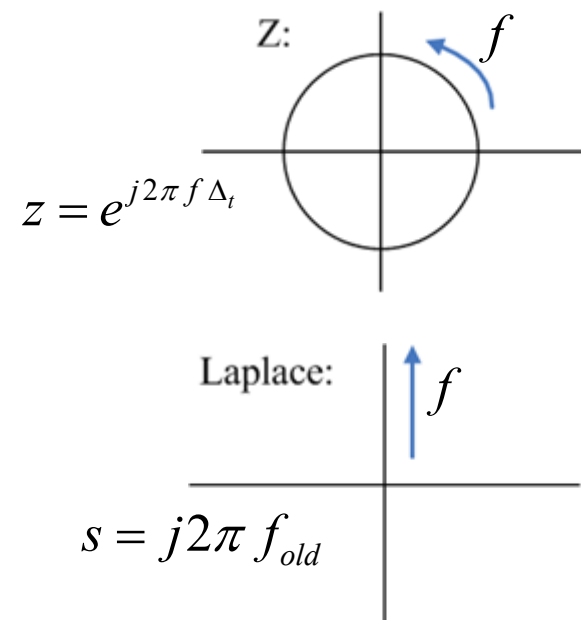
代入

参考page 25、page26

$$\begin{aligned} j2\pi f_{old} &= c \frac{1 - e^{-j2\pi f_{new} \Delta_t}}{1 + e^{-j2\pi f_{new} \Delta_t}} = c \frac{e^{j\pi f_{new} \Delta_t} - e^{-j\pi f_{new} \Delta_t}}{e^{j\pi f_{new} \Delta_t} + e^{-j\pi f_{new} \Delta_t}} \\ &= c \frac{j \sin(\pi f_{new} \Delta_t)}{\cos(\pi f_{new} \Delta_t)} \end{aligned}$$

$$2\pi f_{old} = c \tan(\pi f_{new} \Delta_t)$$

$$f_{new} = \frac{1}{\pi \Delta_t} \operatorname{atan}\left(\frac{2\pi}{c} f_{old}\right) = \frac{f_s}{\pi} \operatorname{atan}\left(\frac{2\pi}{c} f_{old}\right)$$



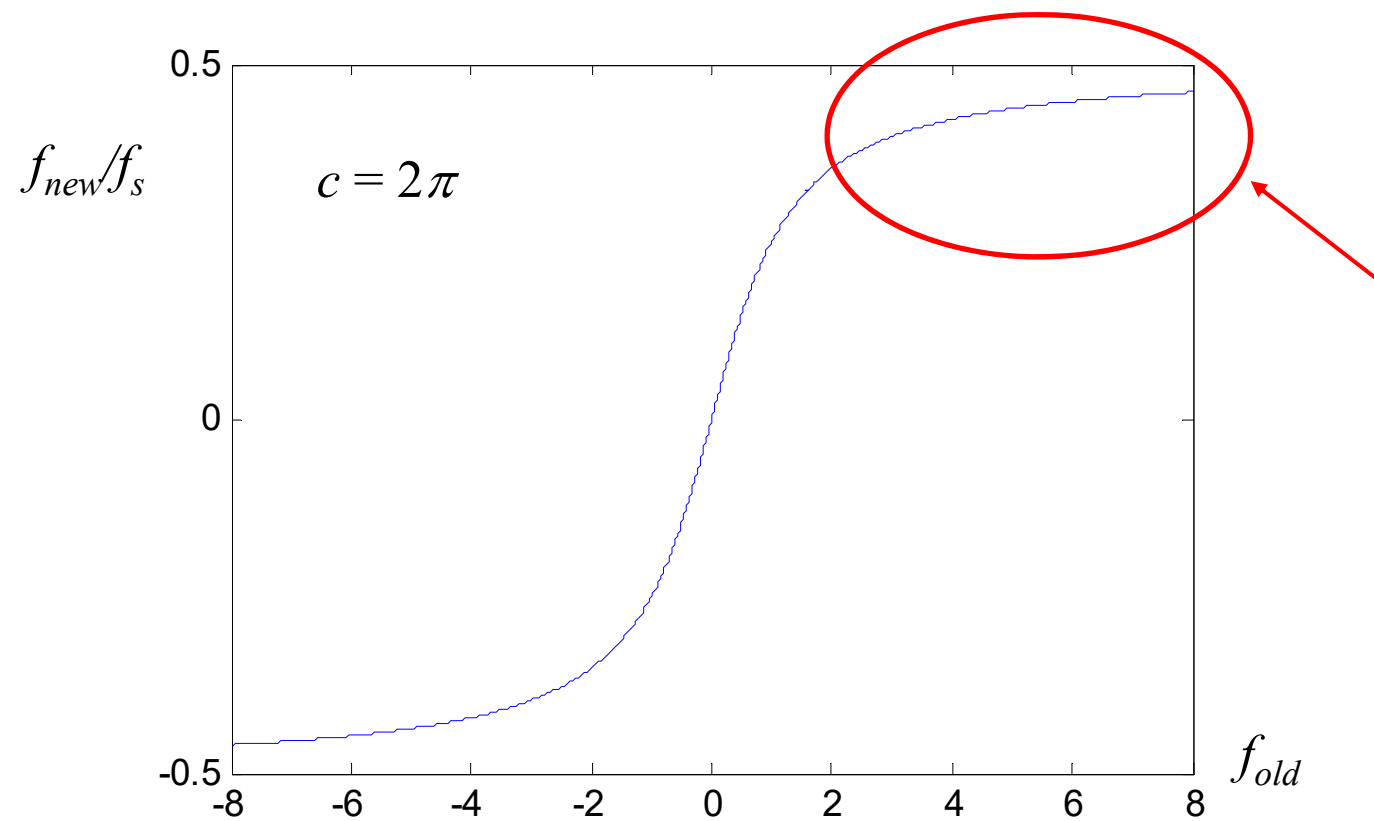
- Suppose that the Laplace transform of the analog filter $h_a(t)$ is $H_{a,L}(s)$

The Z transform of the digital filter $h[n]$ is $H_z(z)$

$$H_z(z) = H_{a,L}\left(c \frac{1 - z^{-1}}{1 + z^{-1}}\right)$$

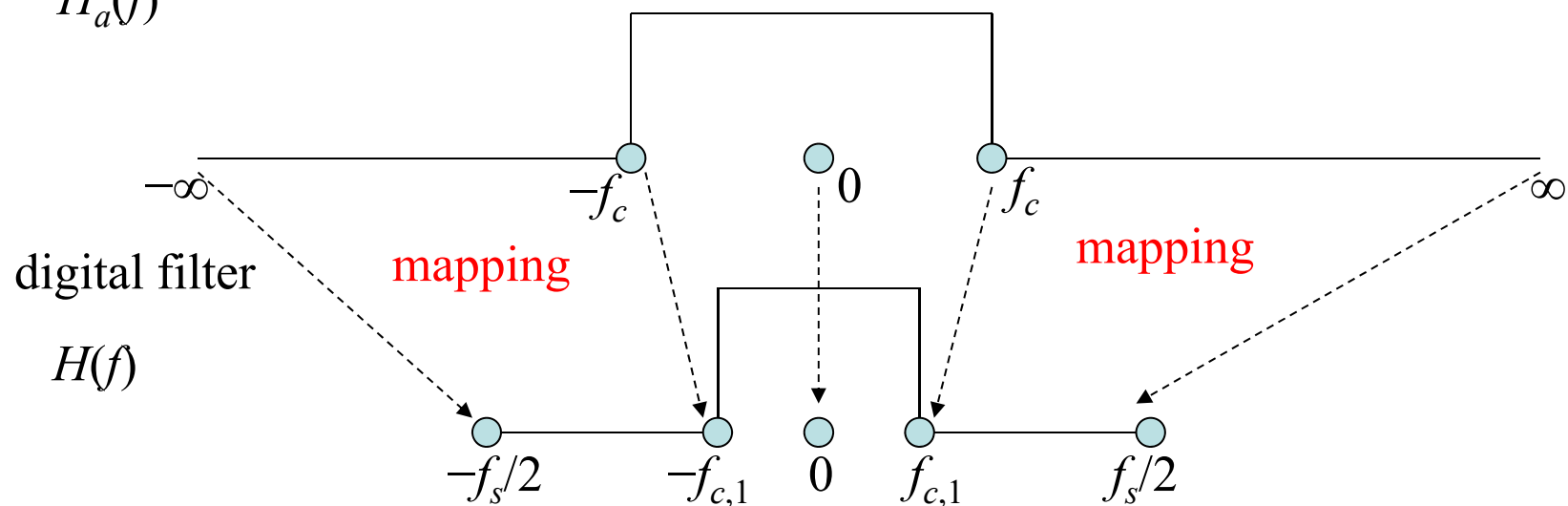
$$f_{new} = \frac{f_s}{\pi} \operatorname{atan} \left(\frac{2\pi}{c} f_{old} \right)$$

f_{old}	$-\infty$	0	∞	1
f_{new}				



analog filter

$$H_a(f)$$



$$f_{c,1} = \frac{f_s}{\pi} \operatorname{atan} \left(\frac{2\pi}{c} f_c \right)$$

Advantage of the bilinear transform

Disadvantage of the bilinear transform

附錄一：學習 DSP 知識把握的要點

- (1) **Concepts**: 這個方法的核心概念、基本精神是什麼
- (2) **Comparison**: 這方法和其他方法之間，有什麼相同的地方？
有什麼相異的地方
- (3) **Advantages**: 這方法的優點是什麼
(3-1) Why? 造成這些優點的原因是什麼
- (4) **Disadvantages**: 這方法的缺點是什麼
(4-1) Why? 造成這些缺點的原因是什麼
- (5) **Applications**: 這個方法要用來處理什麼問題，有什麼應用
- (6) **Innovations**: 這方法有什麼可以改進的地方
或是可以推廣到什麼地方