

◎ 2-H Relations among Filter Length N , Transition Band, and Accuracy

◆ Suppose that we want:

- ① passband ripple $\leq \delta_1$,
- ② stopband ripple $\leq \delta_2$,
- ③ width of transition band $\leq \Delta F$ (expressed by **normalized frequency**)

$$\Delta F = (f_1 - f_2)/f_s = (f_1 - f_2)T \quad (f_s: \text{sampling frequency, } T: \text{sampling interval})$$

Then, the estimated length N of the digital filter is:

$$N = \frac{2}{3} \frac{1}{\Delta F} \log_{10} \left(\frac{1}{10\delta_1\delta_2} \right)$$

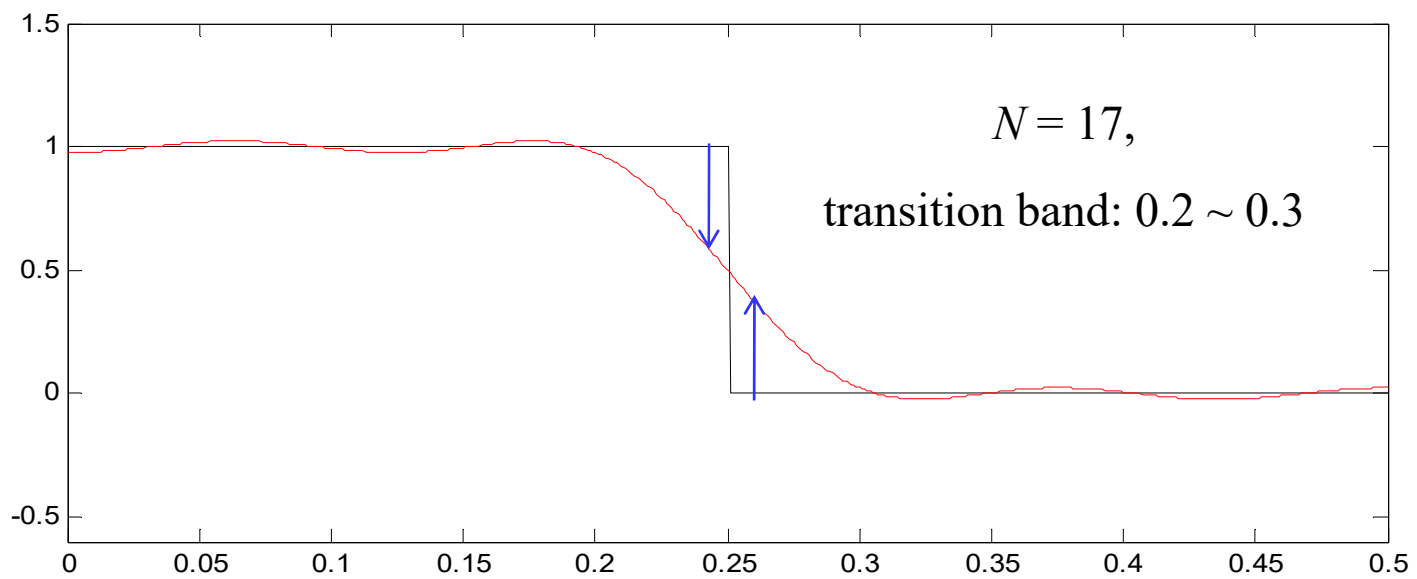
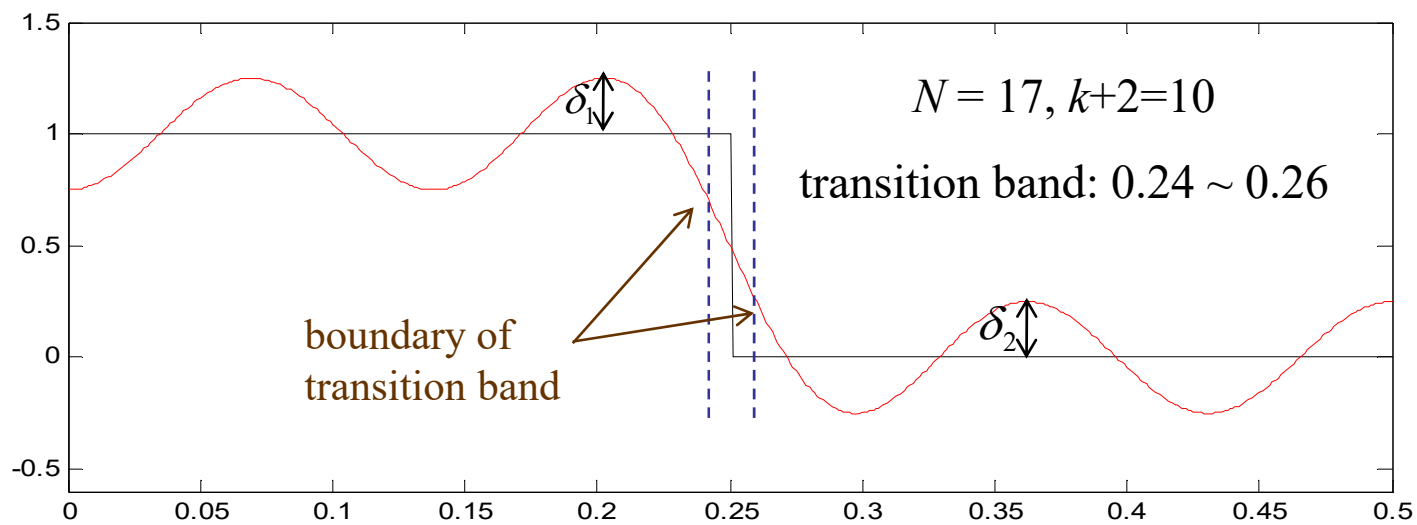
- When there are two transition bands, $\Delta F = \min(\Delta F_1, \Delta F_2)$
- 犧牲 transition band 的 frequency response, 換取較高的 passband and stopband accuracies

$$N = \frac{2}{3} \frac{1}{\Delta F} \log_{10} \left(\frac{1}{10\delta_1\delta_2} \right) \quad \frac{3}{2} N \Delta F = \log_{10} \left(\frac{1}{10\delta_1\delta_2} \right)$$

$$\delta_1\delta_2 = 10^{-3N\Delta F/2-1}$$

[Ref] F. Mintzer and L. Bede, "Practical design rules for optimum FIR bandpass digital filter", *IEEE Trans. ASSP*, vol. 27, no. 2, pp. 204-206, Apr. 1979.

問題：假設 $\sqrt{10}\delta_1 = \sqrt{10}\delta_2 = \delta$ ， N 為固定，
當 ΔF 變為 A 倍時， δ 變為多少？

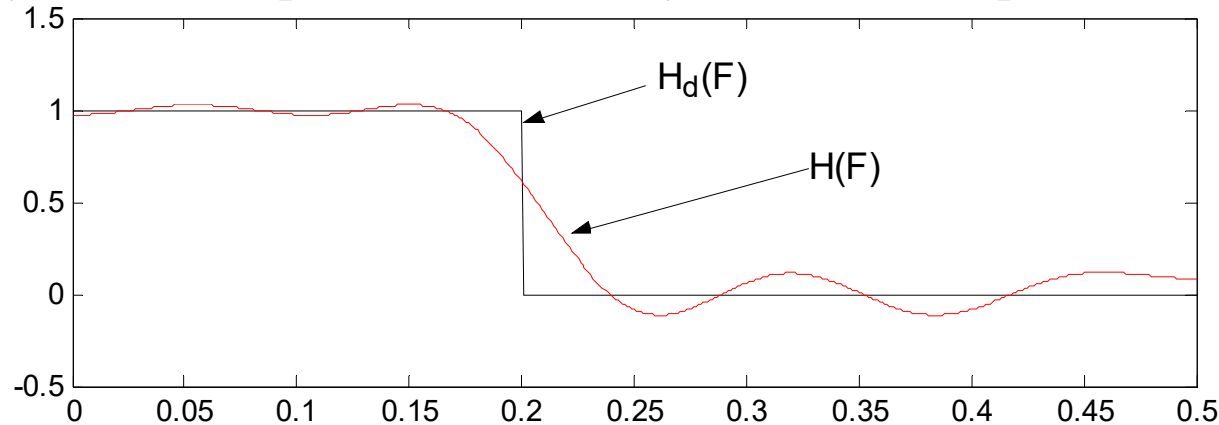


© 2-I Relations between Weight Functions and Accuracy

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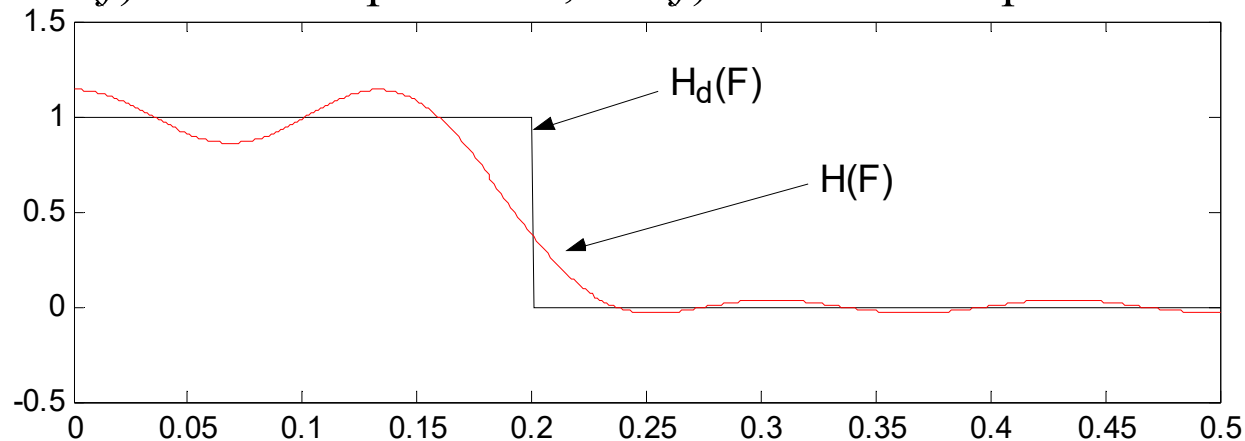
If we treat the passband more important than the stop band

$W(f) = 1$ in the passband, $0 < W(f) < 1$ in the stopband

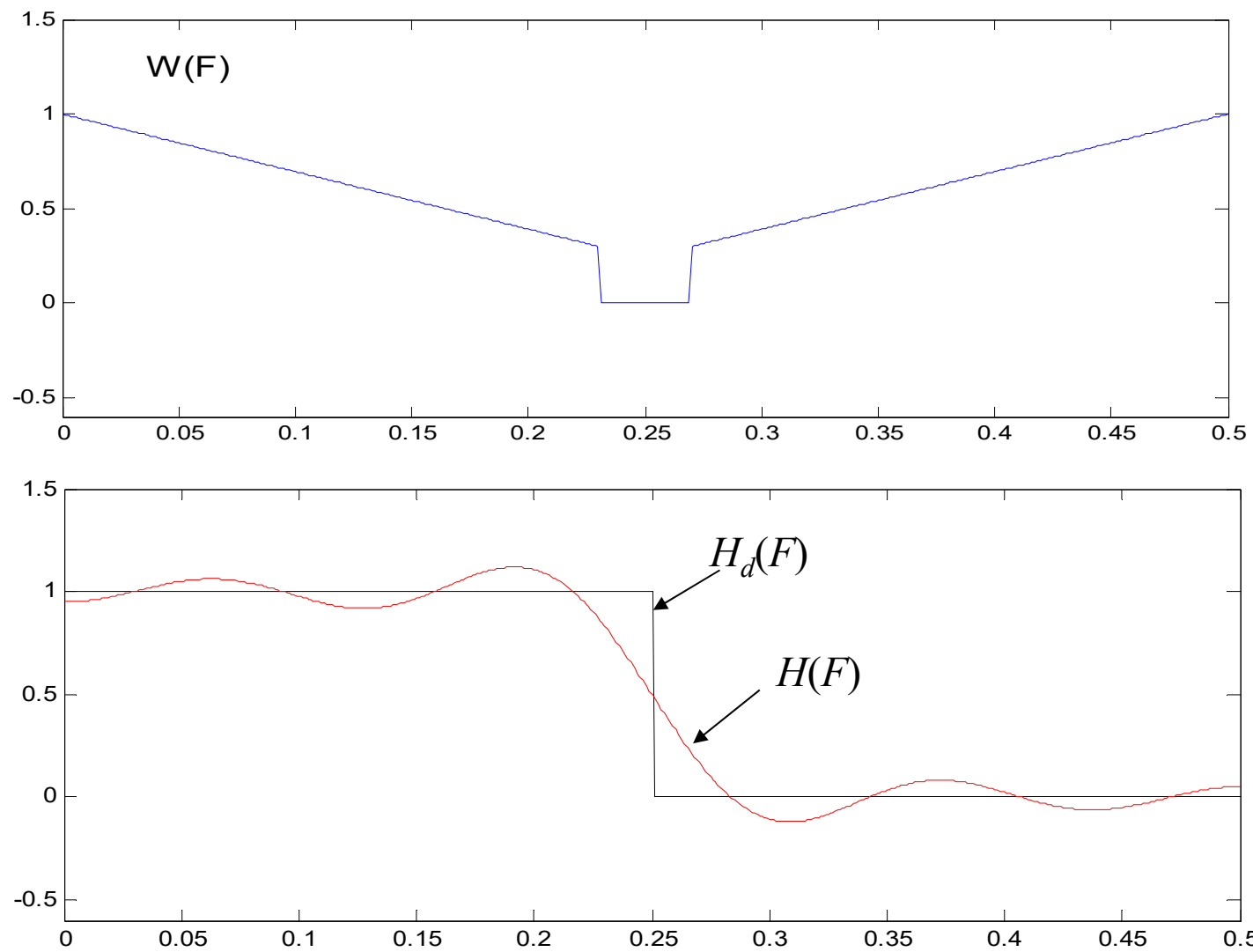


If we treat the stop band more important than the pass band

$0 < W(f) < 1$ in the passband, $W(f) = 1$ in the stopband



Larger error near the transition band



◎ 2-J FIR Filter in MSE Sense with Weight Functions

$$R(F) = \sum_{n=0}^k s[n] \cos(2\pi n F) \quad \text{可對照 pages 49~51}$$

$$MSE = \int_{-1/2}^{1/2} W(F) |R(F) - H_d(F)|^2 dF \quad F = f/f_s$$

$$= \int_{-1/2}^{1/2} W(F) \left(\sum_{\tau=0}^k s[\tau] \cos(2\pi \tau F) - H_d(F) \right)^2 dF$$

$$\frac{\partial MSE}{\partial s[n]} = 2 \int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \left(\sum_{\tau=0}^k s[\tau] \cos(2\pi \tau F) - H_d(F) \right) dF = 0$$

$$\sum_{\tau=0}^k s[\tau] \int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \cos(2\pi \tau F) dF - \int_{-1/2}^{1/2} W(F) H_d(F) \cos(2\pi n F) dF = 0$$

$$n = 0 \sim k$$

問題： $\int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \cos(2\pi \tau F) dF \neq 0$ when $n \neq \tau$

(not orthogonal)

$$\sum_{\tau=0}^k s[\tau] \int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \cos(2\pi \tau F) dF = \int_{-1/2}^{1/2} W(F) H_d(F) \cos(2\pi n F) dF$$

$\tau = 0 \sim k, n = 0 \sim k$

可以表示成 $(k+1) \times (k+1)$ matrix operation

$$\begin{array}{l} \mathbf{n} = \mathbf{0} \\ \mathbf{n} = \mathbf{1} \\ \mathbf{n} = \mathbf{2} \\ \vdots \\ \mathbf{n} = \mathbf{k} \end{array} \begin{array}{ccccc} \mathbf{\tau} = \mathbf{0} & \mathbf{\tau} = \mathbf{1} & \mathbf{\tau} = \mathbf{2} & \cdots & \mathbf{\tau} = \mathbf{k} \\ \left[\begin{array}{ccccc} B[0,0] & B[0,1] & B[0,2] & \cdots & B[0,k] \\ B[1,0] & B[1,1] & B[1,2] & \cdots & B[1,k] \\ B[2,0] & B[2,1] & B[2,2] & \cdots & B[2,k] \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ B[k,0] & B[k,1] & B[k,2] & \cdots & B[k,k] \end{array} \right] \left[\begin{array}{c} s[0] \\ s[1] \\ s[2] \\ \vdots \\ s[k] \end{array} \right] = \left[\begin{array}{c} C[0] \\ C[1] \\ C[2] \\ \vdots \\ C[k] \end{array} \right] \end{array}$$

B**S = C****S = B⁻¹C**

$$B[n, \tau] = \int_{-1/2}^{1/2} W(F) \cos(2\pi n F) \cos(2\pi \tau F) dF$$

$$C[n] = \int_{-1/2}^{1/2} W(F) H_d(F) \cos(2\pi n F) dF$$

Q : Is it possible to apply the **transition band** to the FIR filter in the **MSE sense**?

$$MSE = ?$$

$$B[n, \tau] = ?$$

◎ 2-K Four Types of FIR Filter

$$h[n] = 0 \text{ for } n < 0 \text{ and } n \geq N$$

點數為 N

$$H(F) = \sum_{n=0}^{N-1} h[n] \exp(-j2\pi n F)$$

- Type 1 $R(F) = \sum_{n=0}^k s[n] \cos(2\pi n F)$ ← 之前的方法只討論到 Type 1

$$h[n_1] = h[n_2 - n] \text{ and } N \text{ is odd.}$$

(even symmetric)

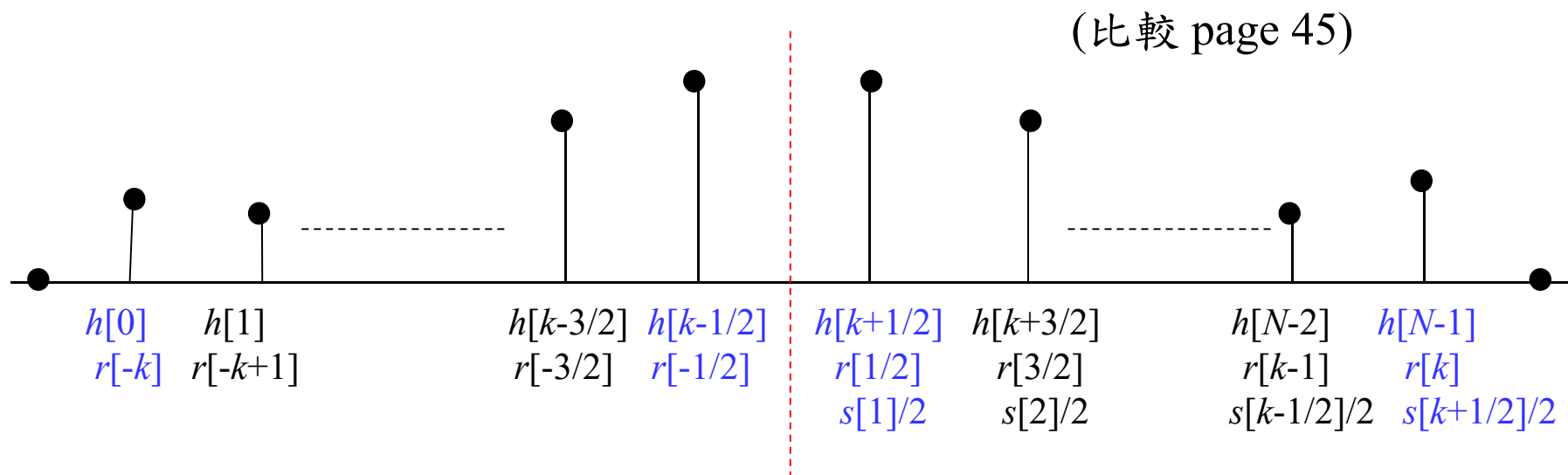
$$k = (N-1)/2$$

- Type 1: $R(F) = \sum_{n=0}^k s[n] \cos(2\pi n F)$
 $\underline{h[n] = h[N-1-n]}$ (even symmetric) and N is odd.
- Type 2: $R(F) = \sum_{n=1}^{k+1/2} s[n] \cos(2\pi (n-1/2) F)$
 $\underline{h[n] = h[N-1-n]}$ (even symmetric) and N is even.
- Type 3: $R(F) = \sum_{n=1}^k s[n] \sin(2\pi n F)$
 $\underline{h[n] = -h[N-1-n]}$ (odd symmetric) and N is odd.
- Type 4: $R(F) = \sum_{n=1}^{k+1/2} s[n] \sin(2\pi (n-1/2) F)$
 $\underline{h[n] = -h[N-1-n]}$ (odd symmetric) and N is even.

$$k = (N-1)/2$$

- Type 2: When $h[n] = h[N-1-n]$ and N is even:
(even symmetric)

令 $r[n] = h[n+k]$, where $k = (N-1)/2$ (注意此時 k 不為整數)



當 $R(F) = \sum_{n=-k}^k r[n] \exp(-j2\pi n F)$

$$R(F) = e^{j2\pi F k} H(F)$$

$$\begin{aligned} R(F) &= \sum_{n=1/2}^k \{r[n] \exp(-j2\pi n F) + r[-n] \exp(j2\pi n F)\} \\ &= \sum_{n=1/2}^k r[n] \{\exp(-j2\pi n F) + \exp(j2\pi n F)\} = \sum_{n=1/2}^k 2r[n] \cos(2\pi n F) \end{aligned}$$

$$R(F) = \sum_{n=1}^{k+1/2} s[n] \cos(2\pi (n-1/2) F)$$

$$n_{(new)} = n_{(old)} + \frac{1}{2} \quad n_{(old)} = n_{(new)} - \frac{1}{2}$$

$$s[n] = 2r[n-1/2] \quad n = 1, 2, \dots, k+1/2$$

設計出 $s[n]$ 之後

$$r[n] = s[n+1/2]/2, \quad h[n] = r[n-k],$$

Design Method for Type 2

$$R(F) = \sum_{n=1}^{k+1/2} s[n] \cos(2\pi(n-1/2)F)$$

由於 n 和 $n+1$ 兩項相加可得

$$\cos(2\pi(n-1/2)F) + \cos(2\pi(n+1/2)F) = 2\cos(\pi F)\cos(2\pi nF)$$

所以可以「判斷」 $R(F)$ 能被改寫成

$$R(F) = \cos(\pi F) \sum_{n=0}^{k_1} s_1[n] \cos(2\pi nF)$$

求 $s_1[n]$ 和 $s[n]$ 之間的關係

$$\begin{aligned} R(F) &= \sum_{n=0}^{k_1} s_1[n] \cos(\pi F) \cos(2\pi nF) \\ &= \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi(n-1/2)F) + \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi(n+1/2)F) \\ &= \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi(n-1/2)F) + \sum_{n=1}^{k_1+1} \frac{1}{2} s_1[n-1] \cos(2\pi(n-1/2)F) \end{aligned}$$

$$R(F) = \sum_{n=0}^{k_1} \frac{1}{2} s_1[n] \cos(2\pi(n-1/2)F) + \sum_{n=1}^{k_1+1} \frac{1}{2} s_1[n-1] \cos(2\pi(n-1/2)F)$$

$$R(F) = \frac{1}{2} s_1[0] \cos(\pi F) + \sum_{n=1}^{k_1} \frac{1}{2} (s_1[n] + s_1[n-1]) \cos(2\pi(n-1/2)F) + \frac{1}{2} s_1[k_1] \cos(2\pi(k_1+1/2)F)$$

$$R(F) = \left(s_1[0] + \frac{1}{2} s_1[1] \right) \cos(\pi F) + \sum_{n=2}^{k-1/2} \frac{1}{2} (s_1[n] + s_1[n-1]) \cos(2\pi(n-1/2)F) + \frac{1}{2} s_1[k-1/2] \cos(2\pi(k)F)$$

(令 $k_1 + 1/2 = k$)

比較係數可得 $s[1] = s_1[0] + \frac{1}{2} s_1[1]$

$$s[n] = \frac{1}{2} (s_1[n] + s_1[n-1]) \quad \text{for } n = 2, 3, \dots, k-1/2$$

$$s[k+1/2] = \frac{1}{2} s_1[k-1/2]$$

$$\begin{aligned}
err(F) &= [R(F) - H_d(F)]W(F) \\
&= \left[\cos(\pi F) \sum_{n=0}^{k-1/2} s_1[n] \cos(2\pi nF) - H_d(F) \right] W(F) \\
&= \left[\sum_{n=0}^{k-1/2} s_1[n] \cos(2\pi nF) - \sec(\pi F) H_d(F) \right] \cos(\pi F) W(F)
\end{aligned}$$

只需將 pages 58-61 的方法當中， $H_d(F)$ 換成 $\sec(\pi F) H_d(F)$

$$\left[\sum_{n=0}^k s[n] \cos(2\pi nF) - H_d(F) \right] W(F)$$

$W(F)$ 換成 $\cos(\pi F) W(F)$

k 換成 $k - 1/2 = N/2 - 1$

注意 $s_1[n]$ 和 $s[n]$ 之間的關係即可

Design Method for Type 3

$$R(F) = \sum_{n=1}^k s[n] \sin(2\pi n F)$$

由於 $n-1$ 和 $n+1$ 兩項相減可得

$$\sin(2\pi(n+1)F) - \sin(2\pi(n-1)F) = 2\sin(2\pi F)\cos(2\pi n F)$$

所以「判斷」可將 $R(F)$ 改寫為

$$R(F) = \sin(2\pi F) \sum_{n=0}^{k_1} s_1[n] \cos(2\pi n F)$$

求 $s_1[n]$ 和 $s[n]$ 之間的關係

$$\begin{aligned} R(F) &= \frac{1}{2} \sum_{n=0}^{k_1} s_1[n] \sin(2\pi(n+1)F) - \frac{1}{2} \sum_{n=0}^{k_1} s_1[n] \sin(2\pi(n-1)F) \\ &= \frac{1}{2} \sum_{n=0}^{k_1} s_1[n] \sin(2\pi(n+1)F) + \frac{s_1[0]}{2} \sin(2\pi F) - \frac{1}{2} \sum_{n=2}^{k_1} s_1[n] \sin(2\pi(n-1)F) \\ &= \frac{1}{2} \sum_{n=1}^{k_1+1} s_1[n-1] \sin(2\pi n F) + \frac{s_1[0]}{2} \sin(2\pi F) - \frac{1}{2} \sum_{n=1}^{k_1-1} s_1[n+1] \sin(2\pi n F) \end{aligned}$$

$$\begin{aligned}
R(F) &= \frac{s_1[0]}{2} \sin(2\pi F) + \frac{1}{2}(s_1[0] - s_1[2]) \sin(2\pi F) \\
&\quad + \frac{1}{2} \sum_{n=2}^{k_1-1} (s_1[n-1] - s_1[n+1]) \sin(2\pi n F) \\
&\quad + \frac{1}{2} s_1[k_1-1] \sin(2\pi k_1 F) + \frac{1}{2} s_1[k_1] \sin(2\pi(k_1+1) F)
\end{aligned}$$

令 $k_1 = k - 1$, 比較係數可得

$$s[1] = s_1[0] - \frac{1}{2} s_1[2]$$

$$s[n] = \frac{1}{2} s_1[n-1] - \frac{1}{2} s_1[n+1] \quad \text{for } n = 2, 3, \dots, k-2$$

$$s[k-1] = \frac{1}{2} s_1[k-2]$$

$$s[k] = \frac{1}{2} s_1[k-1]$$

$$\begin{aligned}
err(F) &= [R(F) - H_d(F)]W(F) \\
&= \left[\sin(2\pi F) \sum_{n=0}^{k-1} s_1[n] \cos(2\pi nF) - H_d(F) \right] W(F) \\
&= \left[\sum_{n=0}^{k-1} s_1[n] \cos(2\pi nF) - \csc(2\pi F) H_d(F) \right] \sin(2\pi F) W(F)
\end{aligned}$$

將 pages 58-61 的方法當中， $H_d(F)$ 換成 $\csc(2\pi F) H_d(F)$

$W(F)$ 換成 $\sin(2\pi F) W(F)$

k 換成 $k-1$

注意 $s_1[n]$ 和 $s[n]$ 之間的關係即可

(Think) : Design the Method for Type 4

一、各種程式語言寫程式共通的原則

- (1) 能夠不在迴圈內做的運算，則移到迴圈外，以節省運算時間
- (2) 寫一部分即測試，不要全部寫完再測試 (縮小範圍比較容易 debug)
- (3) 先測試簡單的例子，成功後再測試複雜的例子

二、Matlab 寫程式特有的技巧

- (1) 迴圈能避免就儘量避免
- (2) 儘可能使用 Matrix 及 Vector operation

Example: 由 1 加到 100，用 Matlab 一行就可以了

```
sum([1:100])
```

完全不需迴圈

三、一些重要的 Matlab 指令

(1) **function**: 放在第一行，可以將整個程式函式化

(2) **tic, toc**: 計算時間

tic 為開始計時，toc 為顯示時間

(3) **find**: 找尋一個 vector 當中不等於 0 的 entry 的位置

範例： $\text{find}([1\ 0\ 0\ 1]) = [1, 4]$

$\text{find}(\text{abs}([-5:5]) \leq 2) = [4, 5, 6, 7, 8]$

(因為 $\text{abs}([-5:5]) \leq 2 = [0\ 0\ 0\ 1\ 1\ 1\ 1\ 1\ 0\ 0\ 0]$)

(4) **'**: Hermitian (transpose + conjugation)，**.'**: transpose

(5) **imread**: 讀圖，**image, imshow, imagesc**: 將圖顯示出來，

(註：較老的 Matlab 版本 imread 要和 double 並用

$A = \text{double}(\text{imread}('Lena.bmp'));$

(6) **imwrite**: 製做圖檔

(7) `xlsread`: 由 Excel 檔讀取資料

```
A = xlsread('檔名', '工作表名', 範圍);
```

例如

```
A = xlsread('test.xlsx', '工作表1', A1:D50);
```

(8) `xlswrite`: 將資料寫成 Excel 檔

(9) `aviread`: 讀取 video 檔

(10) `dlmread`: 讀取 *.txt 或其他類型檔案的資料

(11) `dlmwrite`: 將資料寫成 *.txt 或其他類型檔案

四、寫 Python 版本程式可能會用到的重要指令

建議必安裝模組

```
pip install numpy
```

```
pip install scipy
```

```
pip install opencv-python
```

```
pip install openpyxl # for Excel files
```

(1) 定義函式：使用def

(2) 計算時間

```
import time
```

```
start_time = time.time() #獲取當前時間
```

```
end_time = time.time()
```

```
total_time = end_time - start_time #計算時間差來得到總執行時間
```

感謝2021年擔任助教的蔡昌廷同學

(3) 讀取圖檔、輸出圖檔(建議使用opencv)

```
import cv2
image = cv2.imread(file_name) #預設color channel為BGR
cv2.imwrite(file_name, image) #需將color channel轉為BGR
```

(4) 尋找array中滿足特定條件的值的位址

(相當於 Matlab 的 find 指令)

```
import numpy as np
a = np.array([0, 1, 2, 3, 4, 5])
index = np.where(a > 3) # 回傳array([4, 5])
print(index)
      (array([4, 5], dtype=int64),)
index[0][0]
      4
index[0][1]
      5
```

```
A1= np.array([[1,3,6],[2,4,5]])
```

```
index = np.where(A1 > 3)
```

```
print(index)
```

```
(array([0, 1, 1], dtype=int64), array([2, 1, 2], dtype=int64))
```

(代表滿足 $A1 > 3$ 的點的位置座標為 [0, 2], [1, 1], [1, 2])

```
[index[0][0], index[1][0]]
```

```
[0, 2]
```

```
[index[0][1], index[1][1]]
```

```
[1, 1]
```

```
[index[0][2], index[1][2]]
```

```
[1, 2]
```

$$\mathbf{A}_1 = \begin{bmatrix} 1 & 3 & 6 \\ 2 & 4 & 5 \end{bmatrix}$$

(5) Hermitian 、 transpose

```
import numpy as np
result = np.conj(matrix.T) # Hermitian
result = matrix.T # transpose
```

(6) 在 Python 當中讀取 Matlab 當中的 mat 檔

```
data = scipy.io.loadmat('***.mat')
y = np.array(data['y']) # 假設 y 是 ***.mat 當中儲存的資料
```

(7) 在 Python 當中讀取 Excel 檔

```
import openpyxl
data = openpyxl.load_workbook('filename')
data1 = data['工作表名']
A = [row for row in data1.values]
A1 = np.array(A)
A1 = np.double(A1) # 資料數值化
```